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#### Abstract

There is a magnificent mathematical gem missing from most numerical analysis curricula. A literature search of numerical analysis and mathematical modeling texts indicates its absence. What is missing from the numerical analysis toolbox and why isn't it there? The missing tool is the Kalman filter. The Kalman filter requires a modest knowledge of statistics. Is this why it is missing from the toolbox? Read and reach your own conclusion whether this piece of mathematics should be part of a numerical analysis or mathematical modeling curriculum.


## 1. Missing From the Numerical Analysis Toolbox

In a numerical analysis toolbox [3, 10] you find such things as Gauss-Jordon, Gauss-Seidel elimination, polynomial interpolation and extrapolation, the trapezoid rule for integration, Runge-Kutta method for ODE's, etc., but the Kalman filter is missing. In [2] it is claimed that "along with the simplex method" the Kalman filter is "probably the single most useful piece of mathematics developed this century." It played an essential role in getting the Viking I and II spacecraft from earth to Mars. It is used in navigation, computer vision, to track objects, etc., anywhere sensors obtain measurements subject to noise. Engineers and scientists have used the Kalman filter for several decades following the seminal paper [4] written by Rudolph Kalman. Yet mathematicians seem to have overlooked it.

Literally hundreds of journal articles have been published since Kalman's original work, extending the filter in numerous directions. If this is such a significant result in engineering and applied mathematics then why doesn't it appear in today's numerical analysis books and curriculum? One drawback is that it requires some probability and statistics, though the prerequisites are modest. Most students taking an applied mathematics or numerical analysis course would probably take a course in probability and statistics anyway. Still, there is no guarantee they do. The purpose of this article is to expose this useful tool to the mathematician and show that

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the necessary probability and statistics required to understand it could easily be incorporated into a numerical analysis course. The little time spent on prerequisites is well worth the reward of providing knowledge of an extremely useful tool students might use later in graduate school or industry.

## 2. Where to Start

One way to introduce a Kalman filter is to show what it does. Consider a couple of examples which are slight variations of [11]. A Kalman filter is an algorithm, in fact, a recursive set of equations. As a comparison, consider a recursive equation to compute the mean of a set of data values obtained from measurements $x_{1}, x_{2}, \ldots, x_{n-1}, x_{n}$ taken periodically over time where $x_{1}$ is the oldest value up to $x_{n}$, the most recent value. Convince yourself that if $\bar{x}_{n}$ is the mean of $n$ measurements then

$$
\bar{x}_{n}=\frac{(n-1) \bar{x}_{n-1}+x_{n}}{n}
$$

for $n \geq 1$ with initial condition $\bar{x}_{0}=0$. Normally we don't think of a mean or average as being calculated recursively. However, presented this way, it serves as a simple example which processes data recursively in a manner similar to a Kalman filter. And why recursive? Because that is what makes the Kalman filter so attractive. When it comes to processing data in realtime recursion translates into efficiency. And why the term filter? Because the average and Kalman algorithms "filter out" unwanted noise, revealing a, hopefully, good estimate of the truth.

## 3. Water Tank

The first example takes measurements of the water level of a water tank. The contents of the tank cannot be seen and measurements are taken to determine water level over time. Assume the water level is constant over a short period when it is measured. Like all sensors or measuring devices they have inherent noise. The Kalman filter has proven to be an indispensable tool to deal with noise.

Before looking into the details of the Kalman filter equations, lets see how it compares to taking the average. Ten measurements are taken. Figure 1 shows the results of repeatedly taking the average of the measurements so far, a sort of running average. The true water level, measurements and running average are plotted in Figure 1. The data was purposely made noisy to see how the running average copes with it.

Figure 2 shows how the Kalman filter deals with the noisy measurements. If you compare Figures 1 and 2, you see little difference. In fact, when both the running average and Kalman filter plots were put in the same figure the


Figure 1. Tank filling using running average.
second wrote over and covered the first. That is why two separate figures were made.

How were these plots produced and what do the Kalman equations look like? The implementation of the Kalman and running average equations, together with the plots were produced with the software package Octave [9]. Octave is very similar to Matlab [6] but Octave is free and open source. The Kalman equations during the $n$th iteration for the water tank example

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Figure 2. Tank filling using Kalman filter
are

$$
\begin{aligned}
x_{n} & =x_{n}^{-} \\
P_{n} & =P_{n}^{-}+Q, \\
K_{n} & =P_{n}^{-}\left(P_{n}^{-}+R\right)^{-1}, \\
x_{n} & =x_{n}^{-}+K_{n}\left(y_{n}-x_{n}^{-}\right), \\
P_{n} & =\left(I-K_{n}\right) P_{n}^{-}, \\
x_{n+1}^{-} & =x_{n}, \\
P_{n+1}^{-} & =P_{n}
\end{aligned}
$$

where $x_{n}^{-}$is the previous state, $P_{n}^{-}$is the previous state variance, $Q$ is the process variance, $R$ is measurement noise, $K_{n}$ is referred to as the Kalman gain, $y_{n}$ is a measurement, $I$ is the identity matrix and $P_{n}$ is used to update $P_{n}^{-}$. I know what you are thinking. What do the previous state variance, process variance, measurement noise, Kalman gain, etc., mean? Please be patient. More is said about them below. Whatever is going on, you must

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admit, somehow the Kalman estimates, like the running average, begin to approach the true water level in spite of the noisy data.

The first two equations are called the prediction equations, the next three the Kalman equations and the last two the update equations. Only the middle three equations are often referred to as the Kalman equations proper. Initially, $x_{0}^{-}=0, P_{0}^{-}=1000$ (high state variance), $R=0.1$, $Q=0.0001$ (small process variance), measurements are taken every $\Delta t=1$ sec , and true water level is 1 .

Upper case letters normally represent matrices and lower case vectors. In the water tank example all quantities are one dimensional (scalars). The exponent -1 on the right side of the equation for $K_{n}$ usually represents the inverse of a matrix but since all quantities here are scalars it denotes the reciprocal of a number. The algorithm repeatedly uses the equations to recalculate as more measurements are processed. The Kalman filter estimates plotted in Figure 2 are the values $x_{n}^{-}$.

## 4. Vehicle Position and Velocity

The next example estimates the position and velocity of a vehicle from simulated noisy sensor data. Figure 3 shows plots of the true position of the vehicle, estimated position obtained from a Kalman filter and running average, and the measurements. As Figure 3 indicates, the Kalman filter clearly out performs the running average. In fact, in spite of how noisy the data is, the Kalman filter is obviously giving a very good estimate of the true position. The running average drifts away from the true position over time. Figure 4 shows the difference between true position minus measurement and true position minus Kalman estimated position, and Figure 5 shows true and estimated velocity. The figures clearly show that whatever is going on here the results look good. There must be something to this Kalman filter.

The moving vehicle can be described by the following linear system

$$
\begin{aligned}
\mathbf{x}_{n+1} & =A \mathbf{x}_{n}+B \mathbf{u}_{n}+\mathbf{w}_{n} & & \text { state equation } \\
\mathbf{y}_{n} & =C \mathbf{x}_{n}+\mathbf{z}_{n} & & \text { output equation }
\end{aligned}
$$

where $A, B$ and $C$ are matrices, $n$ is a time index, vector $\mathbf{x}_{n}$ is the state of the system, vector $\mathbf{u}_{n}$ is a known input to the system, vector $\mathbf{y}_{n}$ is measured output, vector $\mathbf{w}_{n}$ is process noise and vector $\mathbf{z}_{n}$ is measurement noise. The state of the system is the vehicle's position and velocity. In the absence of noise, velocity and position can be updated from time index $n$ to $n+1$,

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Figure 3. Vehicle position, Kalman filter and average estimates.
constituting time interval $\Delta t$, as follows

$$
\begin{aligned}
& v_{n+1}=v_{n}+(\Delta t) u_{n} \\
& p_{n+1}=p_{n}+(\Delta t) v_{n}+\frac{(\Delta t)^{2}}{2} u_{n}
\end{aligned}
$$

where $u_{n}$ is input acceleration and the second equation was obtained from the first by integrating. When noise is taken into consideration the equations for velocity and position become

$$
\begin{aligned}
& v_{n+1}=v_{n}+(\Delta t) u_{n}+v_{n}^{\sim} \\
& p_{n+1}=p_{n}+(\Delta t) v_{n}+\frac{(\Delta t)^{2}}{2} u_{n}+p_{n}^{\sim}
\end{aligned}
$$

where $v_{n}^{\sim}$ and $p_{n}^{\sim}$ are velocity and position noise. The later two equations for velocity and position are more realistic because they take instrument noise into account.


Figure 4. Vehicle position measurement and estimate error

## 5. System State

The state vector is $\mathbf{x}_{n}=\binom{p_{n}}{v_{n}}$ and the linear system can be written as

$$
\begin{aligned}
& \mathbf{x}_{n+1}=\left(\begin{array}{cc}
1 & \Delta t \\
0 & 1
\end{array}\right) \mathbf{x}_{n}+\binom{(\Delta t)^{2} / 2}{T} \mathbf{u}_{n}+\mathbf{w}_{n} \\
& \mathbf{y}_{n+1}=\left(\begin{array}{ll}
0 & 1
\end{array}\right) \mathbf{x}_{n}+\mathbf{z}_{n}
\end{aligned}
$$

## 6. Process and Measurement Noise Covariance

Assume measurements are taken every $\frac{1}{10}$ sec, i.e., $\Delta t=0.1 \mathrm{sec}$. Also assume $p_{0}=0, v_{0}=0$ and acceleration is constant at $1 \mathrm{ft} / \mathrm{sec}^{2}$, that is, $a_{0}=a=1 \mathrm{ft} / \mathrm{sec}^{2}$. Then $p=p_{0}+(\Delta t) v_{0}+\frac{(\Delta t)^{2}}{2} a_{0}=0+(.1)(0)+$ $\frac{(0.1 \mathrm{sec})^{2}}{2} \times \frac{1 \mathrm{ft}}{\mathrm{sec}^{2}}=0.005 \mathrm{ft}$. Assume acceleration noise is $\sigma_{a}=0.2 \mathrm{ft} / \mathrm{sec}^{2}$. The standard deviation of position noise is therefore $\sigma_{P}=(0.005)(0.2)=$ $10^{-3}$. Also $v=v_{0}+(\Delta t) a_{0}=0+(0.1 \mathrm{sec})\left(\frac{1 \mathrm{ft}}{\mathrm{sec}^{2}}\right)=0.1 \mathrm{ft} / \mathrm{sec}$. The


Figure 5. Vehicle velocity true value and Kalman filter
standard deviation of velocity noise is therefore $\sigma_{V}=(0.1)(0.2)=0.02=$ $2 \times 10^{-2}$. This accounts for process noise that originates from acceleration noise. Consequently, the process noise vector $\mathbf{w}=\binom{10^{-3}}{2 \times 10^{-2}}$.

Assume process position noise $P$ is a a random variable (RV) having a normal distribution with mean $\mu_{P}=0$ and variance $\sigma_{P}^{2}$. The statistician's notation for this is $P \sim N\left(0, \sigma_{P}^{2}\right)$, which we adopt here. Assume process velocity noise $V$ is an RV and $V \sim N\left(0, \sigma_{V}^{2}\right)$. So the standard deviation for process position noise is $\sigma_{P}=\frac{(\Delta t)^{2}}{2} a \sigma_{a}$ and the standard deviation for process velocity noise is $\sigma_{V}=(\Delta t) a \sigma_{a}$. The process noise vector is $\mathbf{w}=\binom{\sigma_{P}}{\sigma_{V}}$. Consequently, the process noise covariance matrix is

$$
\begin{aligned}
S_{w} & =\operatorname{Cov}(P, V) \\
& =E\left[\mathbf{w w}^{T}\right] \\
& =E\left[\binom{\sigma_{P}}{\sigma_{V}}\left(\begin{array}{ll}
\sigma_{P} & \sigma_{V}
\end{array}\right)\right] \\
& =\left(\begin{array}{cc}
\sigma_{P}^{2} & \sigma_{P} \sigma_{V} \\
\sigma_{V} \sigma_{P} & \sigma_{V}^{2}
\end{array}\right) \\
& =\left(\begin{array}{cc}
\left(\frac{(\Delta t)^{2}}{2} a \sigma_{a}\right)^{2} & \frac{(\Delta t)^{2}}{2} a \sigma_{a}(\Delta t) a \sigma_{a} \\
(\Delta t) a \sigma_{a} \frac{(\Delta t)^{2}}{2} a \sigma_{a} & \left((\Delta t) a \sigma_{a}\right)^{2}
\end{array}\right) \\
& =a^{2} \sigma_{a}^{2}\left(\begin{array}{cc}
\frac{(\Delta t)^{4}}{\left(\frac{(\Delta t)^{3}}{2}\right.} & \frac{(\Delta t)^{3}}{2}
\end{array}\right) \\
& =(1)^{2}(0.2)^{2}\left(\begin{array}{ll}
\frac{(0.1)^{4}}{4} & \frac{(0.1)^{3}}{2} \\
\frac{(0.1)^{2}}{2} & (0.1)^{2}
\end{array}\right) \\
& =\left(\begin{array}{cc}
10^{-6} & 2 \times 10^{-5} \\
2 \times 10^{-5} & 4 \times 10^{-4}
\end{array}\right) .
\end{aligned}
$$

Assume measurement position noise $P_{m}$ is an RV and $P_{m} \sim N\left(0, \sigma_{P_{m}}^{2}\right)$ where the measurement position noise scalar is $z=\sigma_{P_{m}}=10 \mathrm{ft}$. Consequently, the measurement position covariance matrix $\operatorname{Cov}\left(P_{m}, P_{m}\right)=S_{z}=$ $\left(\sigma_{P_{m}}\right)\left(\sigma_{P_{m}}\right)^{T}=\sigma_{P_{m}}^{2}=10^{2}$.

## 7. Linear System and Kalman Equations

Now consider the state, output and Kalman equations for the $n$th iteration. First, a few preliminaries. Matrices $A, B$, and $C$ are used in the equations below where $A=\left(\begin{array}{cc}1 & \Delta t \\ 0 & 1\end{array}\right), B=\binom{\frac{(\Delta t)^{2}}{2}}{\Delta t}$ and $C=\left(\begin{array}{ll}1 & 0\end{array}\right)$. Matrix $A$ is called the transition matrix and $B$ the input matrix. The initial state vector is $\mathbf{x}_{0}=\binom{0}{0}$. The initial input acceleration, and all subsequent acceleration values are $\mathbf{u}_{0}=\mathbf{u}_{n}=1$, which are scalars in this case. The random process noise vector is $\mathbf{w}_{n}=\left(\frac{\left(\Delta t t^{2} r_{1}\right.}{(\Delta t) r_{2}}\right)$ and the $n$th noise measurement vector (which is actually a scalar) is $\mathbf{z}_{n}=\sigma_{P_{m}} r_{3}$ where $r_{1}$, $r_{2}$ and $r_{3}$ are random numbers. New random numbers are generated during each iteration of the algorithm. All random numbers are produced by
the randn library function available in Octave and Matlab which generates random numbers based on RV $R$ where $R \sim N(0,1)$. The initial process covariance matrix is $P_{0}=S_{w}$. The state, output and Kalman equations are

$$
\begin{aligned}
& \mathbf{x}_{n}=A \mathbf{x}_{n}^{-}+B \mathbf{u}_{n}+\mathbf{w}_{n} \\
& \mathbf{y}_{n}=C \mathbf{x}_{n}+\mathbf{z}_{n}
\end{aligned}
$$

$$
\begin{aligned}
\hat{\mathbf{x}}_{n} & =A \hat{\mathbf{x}}_{n}^{-}+B \mathbf{u}_{n} \\
\mathbf{v} & =\mathbf{y}_{n}-C \hat{\mathbf{x}}_{n} \\
S & =C P_{n} C^{T}+S_{z} \\
K_{n} & =A P_{n} A^{T} S^{-1} \\
\hat{\mathbf{x}}_{n+1} & =\hat{\mathbf{x}}_{n}+K_{n} \mathbf{v} \\
P_{n+1} & =A P_{n} A^{T}-A P_{n} C^{T} S^{-1} C P_{n} A^{T}+S_{w}
\end{aligned}
$$

where $\mathbf{x}_{n}^{-}$is the previous state, $\hat{\mathbf{x}}_{n}^{-}$and $\hat{\mathbf{x}}_{n}$ are the previous and next state estimate, $\mathbf{v}$ is the innovation vector, $S$ is the covariance of innovation, and $K_{n}$ is the Kalman gain. This is the jargon used by engineers that work in the linear system world. The first two equations, state and output equations, form the linear system. The remaining equations are considered the Kalman equations. The last two equations prepare the state estimate and process covariance matrix for the next iteration. Exponent $T$ denotes the transpose of a matrix and exponent -1 is the inverse of a matrix.

Here is a little explanation of the above equations. In the calculation of $\hat{\mathbf{x}}_{n+1}$, the first term $\hat{\mathbf{x}}_{n}$ is the propagated state estimate from $A \hat{\mathbf{x}}_{n}^{-}+B \mathbf{u}_{n}$ in the absence of a measurement. The second term $K_{n} \mathbf{v}$ is a correction term. It represents the amount by which to correct the propagated state estimate. Consider the equation for the Kalman gain $K_{n}$. If measurement noise is large, $S_{z}$ will be large, so $K_{n}$ is small and we don't give much credibility to measurement $\mathbf{y}_{n}$. On the other hand, if measurement noise is small, $S_{z}$ is small, so $K_{n}$ is large and we give more credibility to the measurement.

## 8. Real-Time Embedded Software

Nowhere else does a Kalman filter have greater impact than in real-time processing. The world of embedded micro-controllers containing a microprocessor, together with analog to digital (ADC) sensors, timers, wireless communication, USB connections, etc., is growing exponentially, and prolific in our everyday lives. Cell phones, sensors in automobiles, robots,

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appliances, medical monitoring devices, routers, etc., just to name a few, contain micro-controllers embedded inside them that process measurements in real-time. A Kalman filter can be implemented efficiently in the real-time embedded software to deal with noise inherent in the measuring devices to produce reliable results. The impact on our lives is obvious. This is why knowledge of it is a tremendous asset for a student entering the working world or graduate school.

## 9. Lost at Sea

Graphs illustrating the Kalman filter and seeing how well it performs in the presence of noisy data is all well and good but the mathematician asks "where's the proof?" A complete proof of the Kalman filter is outside the scope of this article. There are many references such as [1] which provide a proof in a general multidimensional context required by the Kalman filter. However, some steps towards quenching the mathematician's thirst for proof can be presented through the following variant of a one dimensional example found in [7].

You are lost at sea at night and have no idea of your location. You use the stars to locate your position. Your location is specified as the one dimensional distance from a fixed reference point. At time $t_{1}$ you determine that your location is $z_{1}$ based on your measurements. However, your measurements have inherent inaccuracies and you assume that the variance of your measurements is $\sigma_{z_{1}}^{2}$. Your friend, a trained navigator, takes an independent fix of your location. Immediately after you, at time $t_{2}$, obtain a more accurate measurement with mean $z_{2}$ and (smaller) variance $\sigma_{z_{2}}^{2}$. You have two measurements of your location. How do you combine them?

## 10. Bayesian Estimation

Assume a prior distribution of $\mathrm{RV} M$ to have a normal distribution with mean $z_{1}$ and variance $\sigma_{z_{1}}^{2}$, that is, $M \sim N\left(z_{1}, \sigma_{z_{1}}^{2}\right)$. Let $Z_{2}$ be a normal RV for the sample measurement with mean $z_{2}$ and variance $\sigma_{z_{2}}^{2}$, i.e., $Z_{2} \sim N\left(z_{2}, \sigma_{z_{2}}^{2}\right)$. A method [5, 8] known as Bayesian estimation is used to determine the posterior distribution of $M$, given $Z_{2}=z_{2}$, namely

$$
p\left(\mu \mid z_{2}\right)=\frac{h(\mu) f\left(z_{2} \mid \mu\right)}{g\left(z_{2}\right)}
$$

where the probability density function (PDF) of the prior distribution is

$$
h(\mu)=\frac{1}{\sigma_{z_{1}} \sqrt{2 \pi}} e^{-\frac{1}{2}\left(\frac{\mu-z_{1}}{\sigma_{z_{1}}}\right)^{2}} \quad \text { since } \quad M \sim N\left(z_{1}, \sigma_{z_{1}}^{2}\right)
$$

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and, for $M=\mu$, the PDF of the sampling distribution is

$$
f\left(z_{2} \mid \mu\right)=\frac{1}{\sigma_{z_{2}} \sqrt{2 \pi}} e^{-\frac{1}{2}\left(\frac{z_{2}-\mu}{\sigma_{2}}\right)^{2}} \quad \text { since } \quad Z_{2} \sim N\left(z_{2}, \sigma_{z_{2}}^{2}\right)
$$

The marginal PDF $g\left(z_{2}\right)$ will be determined later.
Plug in $h(\mu), f\left(z_{2} \mid \mu\right)$ and $g\left(z_{2}\right)$ to get

$$
p\left(\mu \mid z_{2}\right)=\frac{1}{2 \pi \sigma_{z_{1}} \sigma_{z_{2}} g\left(z_{2}\right)} e^{-\frac{1}{2}\left(\frac{\mu-z_{1}}{\sigma_{z_{1}}}\right)^{2}-\frac{1}{2}\left(\frac{z_{2}-\mu}{\sigma_{z_{2}}}\right)^{2}} .
$$

which looks like a PDF for a normal distribution. It is, indeed, normal. To show this, the exponent (power) of $e$, denoted here as $P(\mu)$, is rewritten so that completing the square can be done.

## 11. Rearrange $P(\mu)$

A little algebra yields

$$
P(\mu)=-\frac{1}{2}\left(\frac{1}{\sigma_{z_{1}}^{2}}+\frac{1}{\sigma_{z_{2}}^{2}}\right) \mu^{2}+\left(\frac{z_{1}}{\sigma_{z_{1}}^{2}}+\frac{z_{2}}{\sigma_{z_{2}}^{2}}\right) \mu-\frac{1}{2}\left(\frac{z_{1}^{2}}{\sigma_{z_{1}}^{2}}+\frac{z_{2}^{2}}{\sigma_{z_{2}}^{2}}\right)
$$

Let $\frac{1}{\sigma^{2}}=\frac{1}{\sigma_{z_{1}}^{2}}+\frac{1}{\sigma_{z_{2}}^{2}}$ so that the coefficient of $\mu^{2}$ in $P(\mu)$ becomes $-\frac{1}{2 \sigma^{2}}$. If $-\frac{1}{2 \sigma^{2}}$ is a common factor of the $\mu^{2}$ and $\mu$ terms of $P(\mu)$, it can be factored out of these terms. Completing the square can then be performed to help identify $p\left(\mu \mid z_{2}\right)$ as the PDF of a normal distribution.

To get $-\frac{1}{2 \sigma^{2}}$ as a factor of $\mu$ solve

$$
\frac{\mu_{1}}{\sigma^{2}}=\frac{z_{1}}{\sigma_{z_{1}}^{2}}+\frac{z_{2}}{\sigma_{z_{2}}^{2}}
$$

for $\mu_{1}$ to get

$$
\mu_{1}=\frac{\sigma_{z_{2}}^{2} z_{1}}{\sigma_{z_{1}}^{2}}+\frac{\sigma_{z_{1}}^{2} z_{2}}{\sigma_{z_{2}}^{2}}
$$

Consequently

$$
\begin{aligned}
P(\mu) & =-\frac{1}{2 \sigma^{2}} \mu^{2}+\frac{\mu_{1}}{\sigma^{2}} \mu-\frac{1}{2}\left(\frac{z_{1}^{2}}{\sigma_{z_{1}}^{2}}+\frac{z_{2}^{2}}{\sigma_{z_{2}}^{2}}\right) \\
& =-\frac{1}{2 \sigma^{2}}\left(\mu^{2}-2 \mu_{1} \mu\right)-\frac{1}{2}\left(\frac{z_{1}^{2}}{\sigma_{z_{1}}^{2}}+\frac{z_{2}^{2}}{\sigma_{z_{2}}^{2}}\right)
\end{aligned}
$$

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## 12. Complete the Square

Complete the square to get

$$
\begin{aligned}
P(\mu) & =-\frac{1}{2 \sigma^{2}}\left(\mu-\mu_{1}\right)^{2}+\frac{\mu_{1}^{2}}{2 \sigma^{2}}-\frac{1}{2}\left(\frac{z_{1}^{2}}{\sigma_{z_{1}}^{2}}+\frac{z_{2}^{2}}{\sigma_{z_{2}}^{2}}\right) \\
& =-\frac{1}{2}\left(\frac{\mu-\mu_{1}}{\sigma}\right)^{2}+R
\end{aligned}
$$

where $R=\frac{\mu_{1}^{2}}{2 \sigma^{2}}-\frac{1}{2}\left(\frac{z_{1}^{2}}{\sigma_{z_{1}}^{2}}+\frac{z_{2}^{2}}{\sigma_{z_{2}}^{2}}\right)$. Function $p\left(\mu \mid z_{2}\right)$ can now be written as

$$
p\left(\mu \mid z_{2}\right)=\frac{e^{R}}{2 \pi \sigma_{z_{1}} \sigma_{z_{2}} g\left(z_{2}\right)} e^{-\frac{1}{2}\left(\frac{\mu-\mu_{1}}{\sigma}\right)^{2}}
$$

The marginal PDF $g\left(z_{2}\right)$ is defined as

$$
\begin{aligned}
g\left(z_{2}\right) & =\int_{\text {range of } M} h(\mu) f\left(z_{2} \mid \mu\right) d \mu \\
& =\int_{-\infty}^{\infty} h(\mu) f\left(z_{2} \mid \mu\right) d \mu \\
& =\int_{-\infty}^{\infty} \frac{e^{R}}{2 \pi \sigma_{z_{1}} \sigma_{z_{2}}} e^{-\frac{1}{2}\left(\frac{\mu-\mu_{1}}{\sigma}\right)^{2}} d \mu \\
& =\frac{e^{R} \sigma \sqrt{2 \pi}}{2 \pi \sigma_{z_{1}} \sigma_{z_{2}}} \int_{-\infty}^{\infty} \frac{1}{\sigma \sqrt{2 \pi}} e^{-\frac{1}{2}\left(\frac{\mu-\mu_{1}}{\sigma}\right)^{2}} d \mu \\
& =\frac{\sigma e^{R}}{\sigma_{z_{1}} \sigma_{z_{2}} \sqrt{2 \pi}}
\end{aligned}
$$

Plug $g\left(z_{2}\right)=\frac{\sigma e^{R}}{\sigma_{z_{1}} \sigma_{z_{2}} \sqrt{2 \pi}}$ into the equation for $p\left(\mu \mid z_{2}\right)$ to get

$$
p\left(\mu \mid z_{2}\right)=\frac{1}{\sigma \sqrt{2 \pi}} e^{-\frac{1}{2}\left(\frac{\mu-\mu_{1}}{\sigma}\right)^{2}}
$$

This is the PDF of a normal distribution with mean $\mu_{1}$ and variance $\sigma^{2}$ where

$$
\mu_{1}=\frac{\sigma_{z_{2}}^{2} z_{1}}{\sigma_{z_{1}}^{2}}+\frac{\sigma_{z_{1}}^{2} z_{2}}{\sigma_{z_{2}}^{2}} \quad \text { and } \quad \frac{1}{\sigma^{2}}=\frac{1}{\sigma_{z_{1}}^{2}}+\frac{1}{\sigma_{z_{2}}^{2}}
$$

## 13. Your Turn

The above derivation of $\mu_{1}$ and $\sigma$ was based on Bayesian estimation. Another approach [12] is to use $\chi^{2}$ estimation. An outline of this approach is given and you are asked to fill in the details. Perhaps this can be the

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start of a capstone or undergraduate research project that you, the student, would like to pursue. Start by considering the following expression

$$
\chi^{2}=\frac{\left(z_{1}-\hat{x}\right)^{2}}{\sigma_{z_{1}}^{2}}+\frac{\left(z_{2}-\hat{x}\right)^{2}}{\sigma_{z_{2}}^{2}}
$$

where $z_{1}, z_{2}, \sigma_{z_{1}}$ and $\sigma_{z_{2}}$ are all the same as before. Treat $\chi^{2}$ as a function of $\hat{x}$ and take the derivative with respect to $\hat{x}$. Set the derivative equal to zero and solve for $\hat{x}$. Does it look familiar?

If $\sigma_{z_{1}}^{2}=E\left[\left(z_{1}-\mu_{x}\right)^{2}\right]$ and $\sigma_{z_{2}}^{2}=E\left[\left(z_{2}-\mu_{x}\right)^{2}\right]$ where $E$ is expected value, find $\sigma_{x}^{2}=E\left[\left(\hat{x}-\mu_{x}\right)^{2}\right]$ by plugging in the expression you obtained for $\hat{x}$. Keep in mind that $z_{1}$ and $z_{2}$ are independent. What do you get for $\sigma_{x}^{2}$ ?

## 14. ObSERVATIONS OF $\mu_{1}$ AND $\sigma^{2}$

Note that $\sigma^{2}<\min \left(\sigma_{z_{1}}^{2}, \sigma_{z_{2}}^{2}\right)$. Therefore, two measurements yield a PDF with smaller variance, and hence, a better estimate of the true position than either single measurement provides. Suppose $\sigma_{z_{2}}^{2}=\sigma_{z_{1}}^{2}$. Use the equation for $\mu_{1}$ to show that $\mu_{1}=\frac{z_{1}+z_{2}}{2}$ which is the average of the two measurements. Suppose $\sigma_{z_{2}}^{2}>\sigma_{z_{1}}^{2}$. Then the coefficient of $z_{1}$ in the expression for $\mu_{1}$ is larger than the coefficient of $z_{2}$ giving more weight to the estimate $z_{1}$ when calculating $\mu_{1}$. This is desirable because $z_{1}$ would be the better estimate since it has a smaller variance. Based on both measurements $z_{1}$ and $z_{2}$ an estimate of $\mu_{1}$ was obtained. As it turns out, it is the maximum likelihood estimate, the weighted least squares estimate and the linear estimate whose variance is less than that of any other linear unbiased estimate. Consequently, it is the "best" estimate according to any reasonable criterion.

## 15. A Little Kalman Filter

The equation for $\mu_{1}$ can be rearranged to derive a formula for the estimate $\hat{x}\left(t_{2}\right)=\mu_{1}$ in terms of the estimate $\hat{x}\left(t_{1}\right)$. Using the fact that $\hat{x}\left(t_{1}\right)=z_{1}$ show that

$$
\hat{x}\left(t_{2}\right)=\hat{x}\left(t_{1}\right)+\frac{\sigma_{z_{1}}^{2}}{\sigma_{z_{1}}^{2}+\sigma_{z_{2}}^{2}}\left[z_{2}-\hat{x}\left(t_{1}\right)\right]
$$

This yields the following Kalman Filter

$$
\hat{x}\left(t_{2}\right)=\hat{x}\left(t_{1}\right)+K\left(t_{2}\right)\left[z_{2}-\hat{x}\left(t_{1}\right)\right]
$$

where

$$
K\left(t_{2}\right)=\frac{\sigma_{z_{1}}^{2}}{\sigma_{z_{1}}^{2}+\sigma_{z_{2}}^{2}}
$$

$K\left(t_{2}\right)$ is the Kalman gain that was mentioned earlier. This equation provides one form of the Kalman filter formula. This gives a way of deriving the
best estimate of your position, $\hat{x}\left(t_{2}\right)$ at time $t_{2}$ based on the best estimate $\hat{x}\left(t_{1}\right)$ at time $t_{1}$ and the measurement $z_{2}$ at time $t_{2}$.

The above Kalman filter equation can be used to derive an analogous formula to calculate the variance $\sigma_{x}^{2}\left(t_{2}\right)$ in terms of $\sigma_{x}^{2}\left(t_{1}\right)$ and $K\left(t_{2}\right)$. Show that

$$
\sigma^{2}=\left[1-K\left(t_{2}\right)\right] \sigma_{z_{1}}^{2}
$$

Since $\sigma_{z_{1}}^{2}=\sigma_{x}^{2}\left(t_{1}\right)$ and $\sigma^{2}=\sigma_{x}^{2}\left(t_{2}\right)$ the last equation can be rewritten as

$$
\sigma_{x}^{2}\left(t_{2}\right)=\left[1-K\left(t_{2}\right)\right] \sigma_{x}^{2}\left(t_{1}\right)
$$

This equation provides a formula for calculating $\sigma_{x}^{2}\left(t_{2}\right)$ from $\sigma_{x}^{2}\left(t_{1}\right)$ and $K\left(t_{2}\right)$ analogous to the equation used to calculate $\hat{x}\left(t_{2}\right)$ from $\hat{x}\left(t_{1}\right)$ and $K\left(t_{2}\right)$. The pair of equations for $\hat{x}\left(t_{2}\right)$ and $\sigma_{x}^{2}\left(t_{2}\right)$ solve what is referred to as the static estimation problem.

## 16. Dynamic Estimation Problem

Returning to the "lost at sea" example, suppose you travel for some time before taking a third measure $z_{3}$ of your position at time $t_{3}$. Assume that the best model of your motion is given by the differential equation

$$
\frac{d x}{d t}=u+w
$$

where $u$ is a nominal velocity and $w$ noise that is assumed to have a normal PDF with zero mean and variance $\sigma_{w}^{2}$. Let $t_{3}^{-}$represent the time just before $t_{3}$ when the third measurement $z_{3}$ is taken.
Given a third measurement $z_{3}$ of your position $x\left(t_{3}\right)$ at $t_{3}$, a fourth PDF having a normal distribution can be defined. The fourth PDF at time $t_{3}^{-}$has a mean $\hat{x}\left(t_{3}^{-}\right)$and variance $\sigma_{x}^{2}\left(t_{3}^{-}\right)$which satisfy the following two equations

$$
\begin{aligned}
\hat{x}\left(t_{3}^{-}\right) & =\hat{x}\left(t_{2}\right)+u\left[t_{3}-t_{2}\right] \\
\sigma_{x}^{2}\left(t_{3}^{-}\right) & =\sigma_{x}^{2}\left(t_{2}\right)+\sigma_{w}^{2}\left[t_{3}-t_{2}\right] .
\end{aligned}
$$

The value of $\hat{x}\left(t_{3}^{-}\right)$is the best estimate of your position at time $t_{3}^{-}$just before measurement $z_{3}$ at time $t_{3}$ is taken and $\sigma_{x}^{2}\left(t_{3}^{-}\right)$is the variance in that estimate. If the Kalman filter equation for $\hat{x}\left(t_{2}\right)$ above is applied with $\hat{x}\left(t_{3}^{-}\right)$being the previous best estimate at time $t_{3}^{-}, z_{3}$ the measurement at time $t_{3}$ we get

$$
\hat{x}\left(t_{3}\right)=\hat{x}\left(t_{3}^{-}\right)+K\left(t_{3}\right)\left[z_{3}-\hat{x}\left(t_{3}^{-}\right)\right]
$$

for the best estimate $\hat{x}\left(t_{3}\right)$ in position at time $t_{3}$ where

$$
K\left(t_{3}\right)=\frac{\sigma_{x}^{2}\left(t_{3}^{-}\right)}{\sigma_{x}^{2}\left(t_{3}^{-}\right)+\sigma_{z_{3}}^{2}}
$$

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Similarly, if the equation above for $\sigma_{x}^{2}\left(t_{2}\right)$ is applied with $\sigma_{x}^{2}\left(t_{3}^{-}\right)$being the best estimate of the variance at time $t_{3}^{-}$, we can calculate the best estimate of the variance $\sigma_{x}^{2}\left(t_{3}\right)$ at time $t_{3}$ as follows

$$
\sigma_{x}^{2}\left(t_{3}\right)=\left[1-K\left(t_{3}\right)\right] \sigma_{x}^{2}\left(t_{3}^{-}\right)
$$

Extending equations $\hat{x}\left(t_{3}^{-}\right)$and $\sigma_{x}^{2}\left(t_{3}^{-}\right)$to vectors yields the general Kalman filter algorithm for updating from one measurement to the next measurement. Extending equations $K\left(t_{3}\right)$ and $\sigma_{x}^{2}\left(t_{3}\right)$ to vectors yields the Kalman filter update at a measurement time.

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